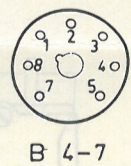
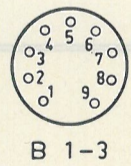
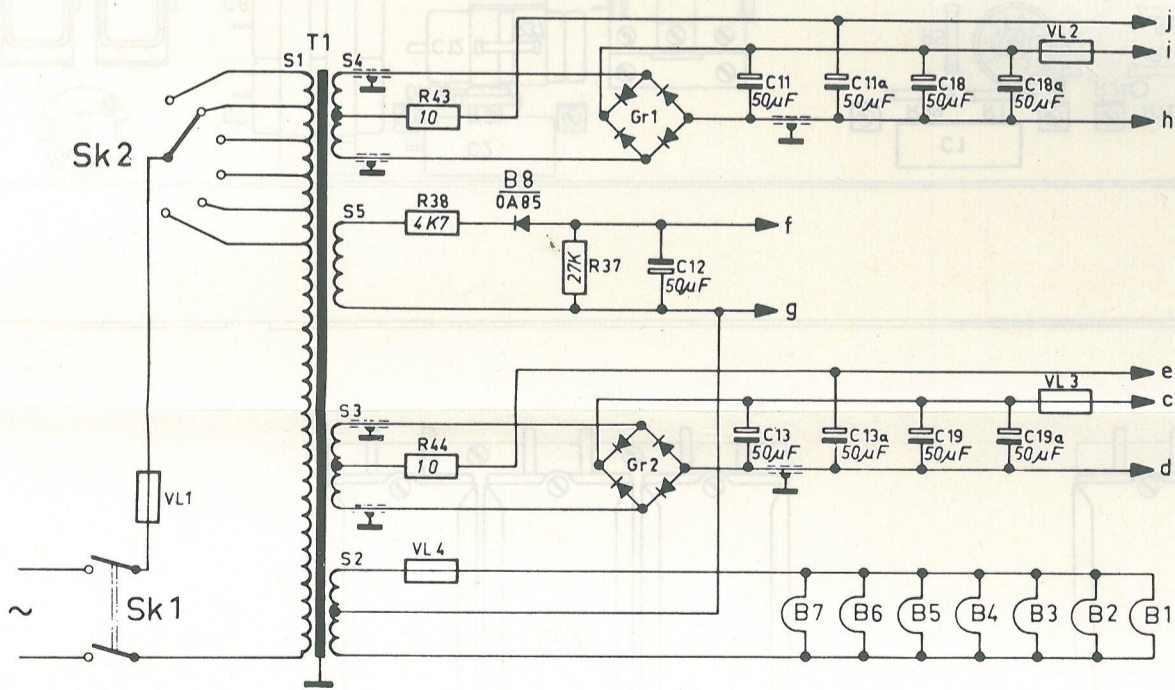
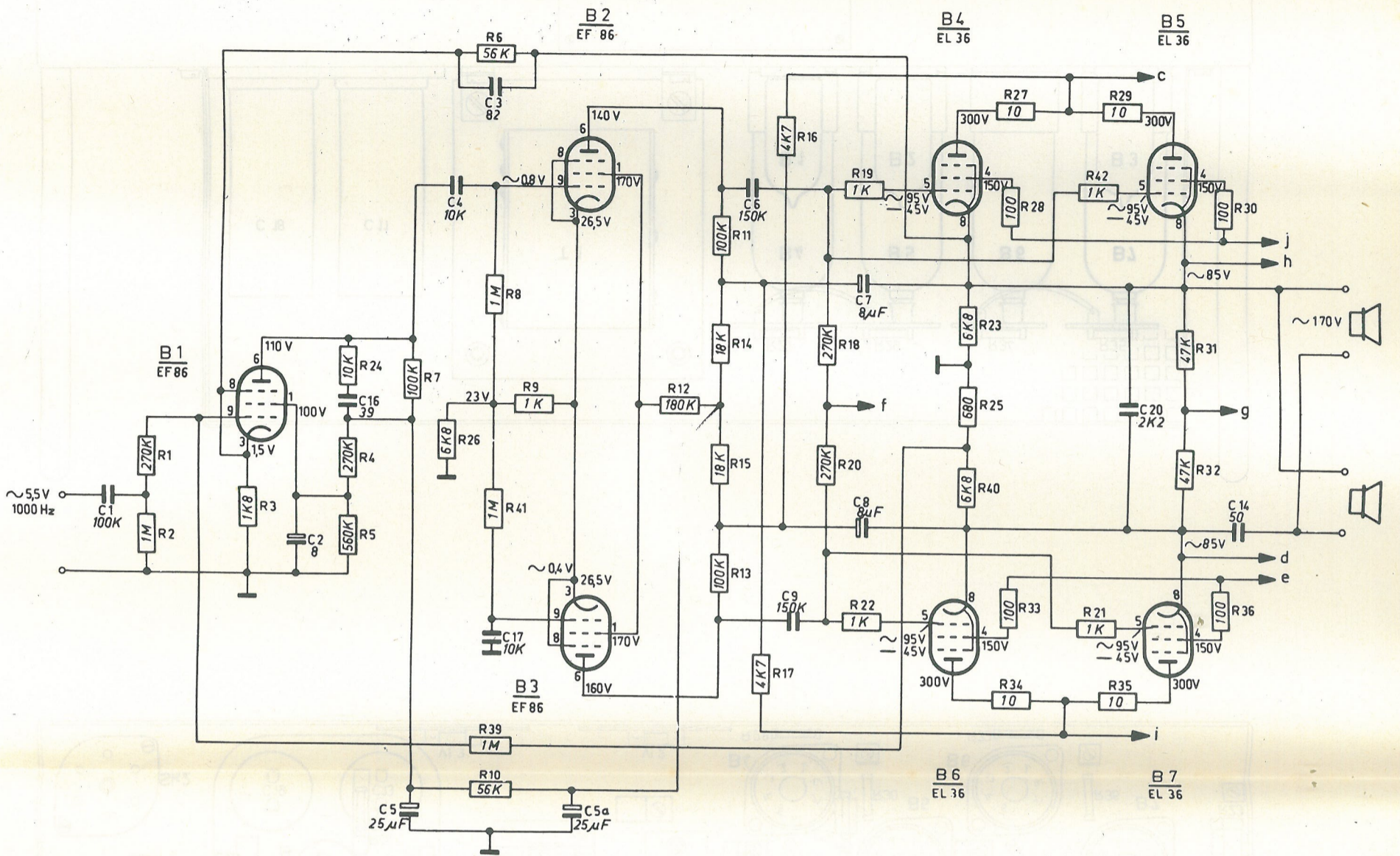


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C	1.	2.	16.	5.	4.	3.17.	12.	5a.11.13.	11a.13a.	18.19.6	9.18a.19a.	7.8	20.	14.																
R	1. 2.	3.	24.4.5.4.3.38.44.7.	26.	6.8.41.39.10.37.9.	12.	11.14.15.13.17.16.	18. 20.19. 22.	23.25.40.27.28.33.34.	42.21.	29.35.	31. 2. 30.36.																		



R 173 01

Fig.1

AG 9007

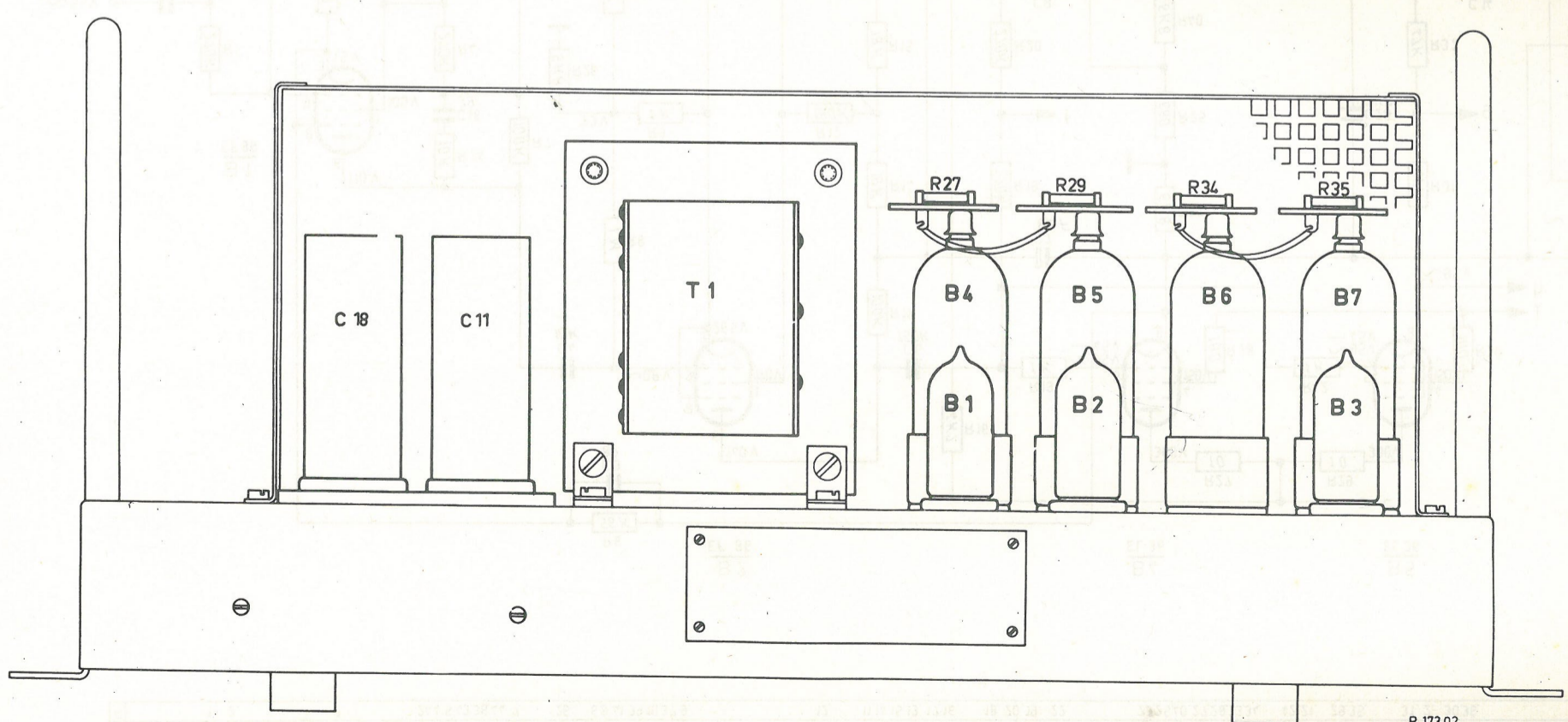
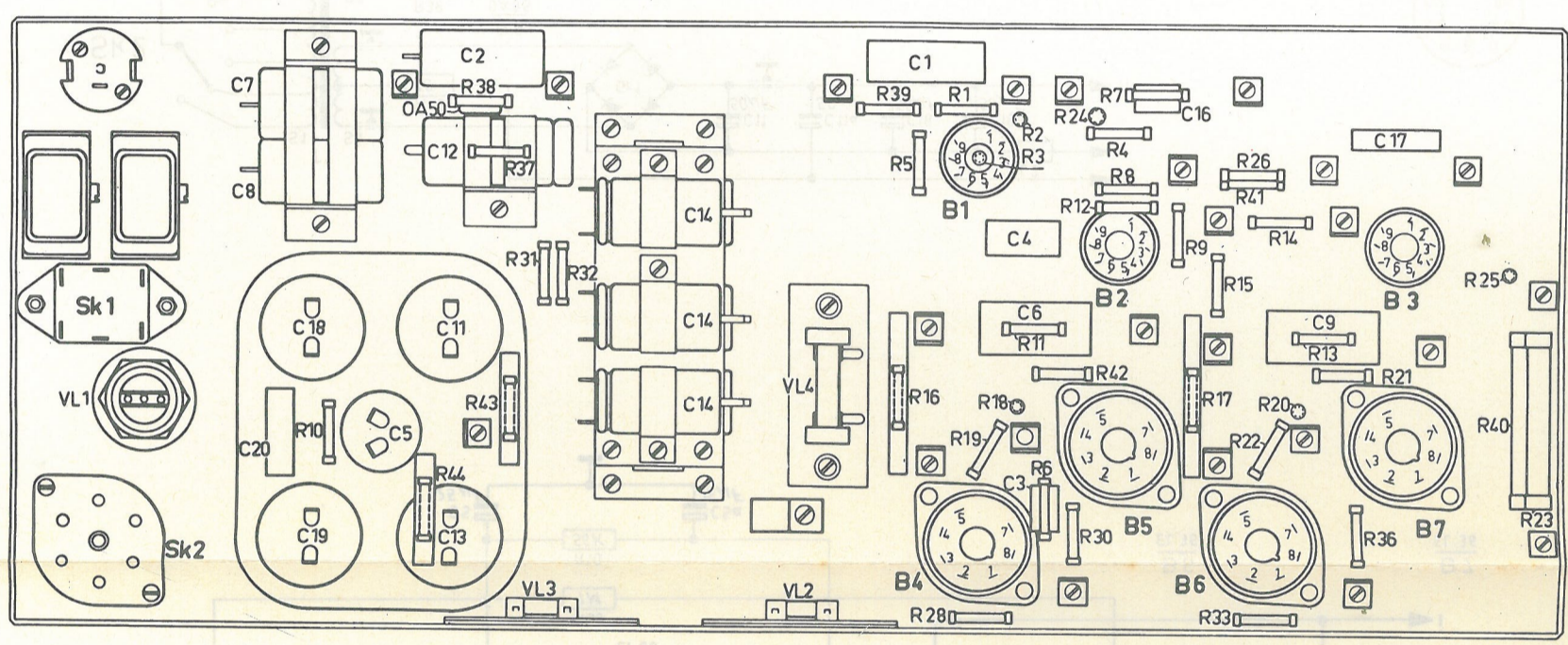
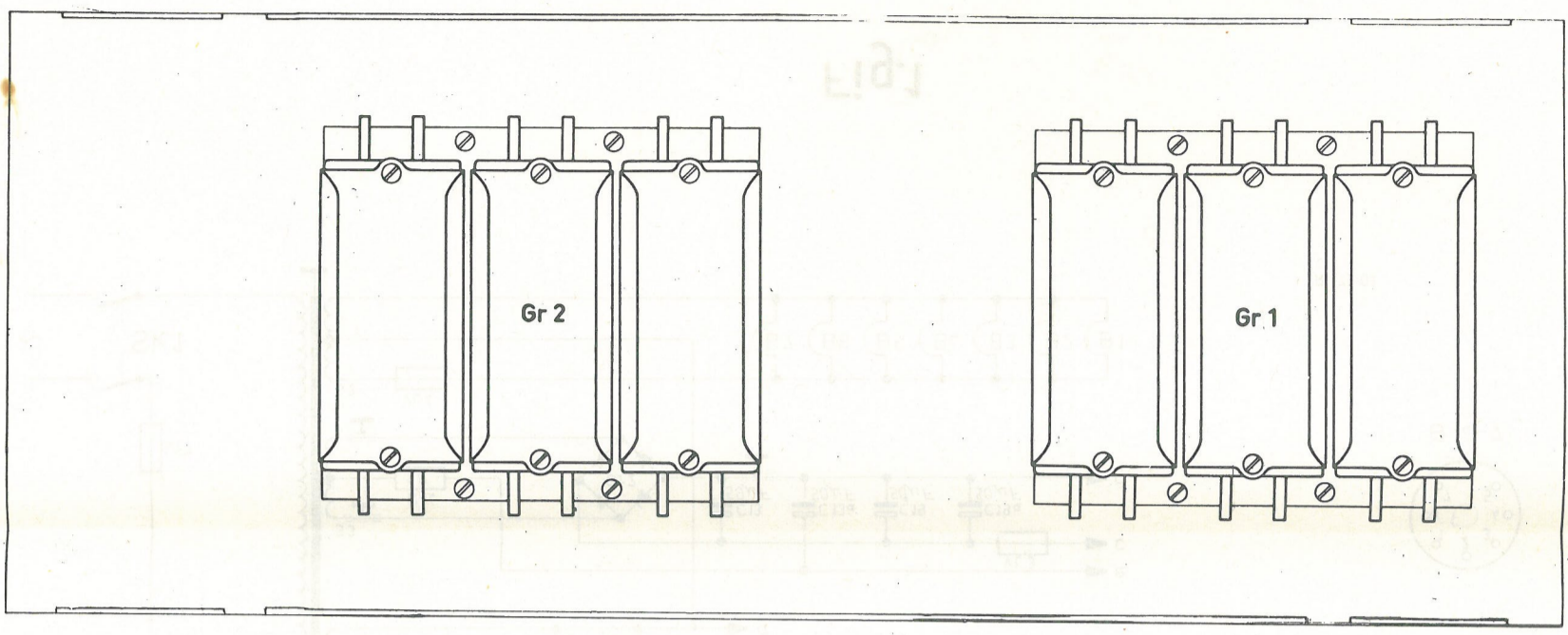


Fig2

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# PHILIPS

## SERVICE NOTES

for the Hi-Fi power amplifier

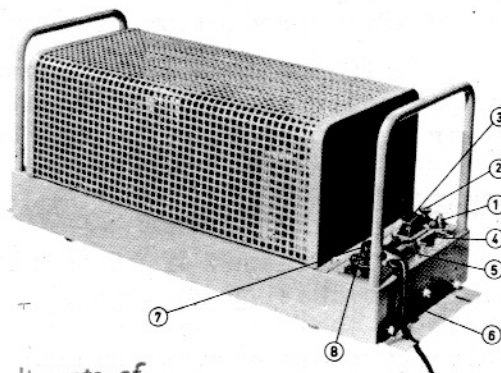
### AG 9007

With the Compliments of...

RADIO WHOLESALERS LTD.

P.O. Box 527

INVERCARGILL



1958. For A.C. mains supply

#### Operation

1. Input socket
2. Earth screw
3. Loudspeaker connection
4. Loudspeaker connection
5. Mains switch
6. Mains flex + plug
7. Fuse holder
8. Voltage adaptor

#### Mains voltages

117-127-145-200-220-245 V  
50 - 100 c/s

#### Power consumption

Unloaded: 85 Watts (220V)  
With 39-W output: 136 W  
With 60-W output: 177 W

#### Output power and distortion

When loaded with 400Ω:  $W_o = 60W$   
When loaded with 800Ω:  $W_c = 39W$

With full output, the intermodulation distortion is smaller than 2%, and the harmonic distortion smaller than 0.3%.

#### Frequency characteristic

This is straight from 25 c/s to 30.000 c/s.

#### Sensitivity

When loaded with 800Ω, the sensitivity amounts to 5.7V with an output capacity of 39 Watts.

When loaded with 400Ω, the sensitivity amounts to 5V with an output capacity of 60 Watts.

#### Hum and Noise Level

Measured with open input:  
-95 dB with respect to 60 Watts output.

#### Dimensions

440 x 215 x 173 mm.

#### Valves

B1, B2, B3 : EF86  
B4, B5, B6, B7 : EL36  
B8 : 0A85  
GR1 : 3x SR300B120  
GR2 : 3x SR300B120  
VL1 : 974/1000 for 220V  
VL2 : 974/250  
VL3 : 974/250  
VL4 : 08 143 02.

93 992 14.1.05

Purpose

This Hi-Fi power amplifier can be used together with the pre-amplifier AG 9004 and one or two loudspeaker combinations AD 5034, for the Hi-Fi reproduction of records, tapes, radio, etc.

If one loudspeaker combination AD 5034 is used, the maximum output power amounts to 39 Watts. With two combinations AD 5034, a power output of 60 Watts can be achieved.

Remark

The output of the amplifier is symmetrical with respect to earth, so that on no account one of the output terminals may be earthed, as in this case half of the output signal is short-circuited. This may cause serious damages of the amplifier.

Measuring of voltages

For this, two resistors of  $390\Omega$  25W are required (code no. B8 300 34B/390E).

These resistors are connected in series to the loudspeaker-sockets, while the junction of the resistors is connected to earth.

All the voltages are measured with respect to earth. For measuring the sensitivity and the step amplification, use can be made of a tone generator, e.g. the GM 2317 or the GM 2315; the measuring frequency chosen is 1000 c/s.

Measuring of the frequency characteristic

The frequency characteristic has to be straight from 20 c/s to 30.000 c/s. Deviation  $\pm 1$  dB.

For this, a tone generator GM 2317 plus a valve voltmeter, e.g. GM 6015 or GM 6017, are needed. The valve voltmeter is connected across one of the load resistors of  $390\Omega$ . After this the same measurement can be carried out across the other load resistor.

Description of the circuit diagram

The signal is fed to g1 of B1 via C1 and R1. C1 has been provided to block any leak current of the cathode follower of the pre-amplifier. R2 functions as a leak resistor.

After having been amplified by B1, the signal is fed to the phase-shifting system, consisting of B2 and B3. Let us explain the working of this system:

Suppose that at a certain moment a positive voltage pulse appears on the grid of B2. In this case, the anode current of B2 will increase, consequently the cathode current increases, which causes an increased voltage across R9. As the grid of B3 for alternating current is connected to earth (via C17), and the cathode receives a positive voltage pulse, the anode current of B3 will temporarily decrease. So we see that the anode currents of B2 and B3 are in opposite phase.

After having been amplified by B2 and B3, the signal is fed to the output valves B4, B5, B6 and B7. B4 and B5 are connected in parallel, and B6 and B7 as well.

The output valves are connected as cathode followers, so that they slightly attenuate the voltage fed, but amplify the energy. The advantage of this circuit is the low output impedance, so that high-

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impedance loudspeakers can be direct connected, and the loudspeaker transformer becomes superfluous.

Another and not less advantage is that a cathode follower practically does not give any distortion. When, moreover, a strong negative feedback is applied for the whole amplifier, the distortion will be negligibly small, which is also the case with the amplifier under discussion.

A disadvantage of the above mentioned circuit is the high grid voltage, which is needed to attain sufficient power. Therefore, the distortion would be shifted from the output valves to the preamplifier. However, this disadvantage has been completely removed in the following way: the relevant output signals have been brought, via C7 and C8, under in the anode circuits of B2 and B3, so that the preamplifier only need give a small voltage, and consequently gives little distortion. The negative grid voltage of the output valves is fixed, and is fed via the point f and g (see circuit diagram).

The output valves are fed by two H.T. units. + High tension from B4 and B5 is fed to point C. - high tension to d. Consequently, the output signal of B4 and B5 appears across R31+R32.

In the same way can be shown that the output signal of B6 and B7 appears across R32+R31. Via R6 and C3, a negative feedback voltage is applied to the cathode of B1. C3 has been provided to suppress any oscillating tendencies. Moreover, part of the output signal is fed via R39 to the grid of B1, which causes a strong negative feedback.

All this helps to keep the amplifier free of distortion.

S1)				R15	18000	Ω	901/18K
S2)				R16	4700	Ω	901/4K7
S3)			V3 617 49	R17	4700	Ω	901/4K7
S4)				R18	0,27	MΩ	901/270K
S5)				R19	1000	Ω	900/1K
C1	0,1	μF	906/100K	R20	0,27	MΩ	900/270K
C2	8	μF	911/P8	R21	1000	Ω	900/1K
C3	82	pF	905/82E	R22	1000	Ω	900/1K
C4	10.000	pF	906/V10K	R23	6800	Ω	2x900/15K *
C5 )	25	μF	913/N25x25	R24	10000	Ω	900/10K
C5a)	25	μF		R25	680	Ω	901/680E
C6	0,15	μF	906/150K	R26	6800	Ω	901/6K8
C7	8	μF	911/P8	R27	10	Ω	901/10E
C8	8	μF	911/P8	R28	100	Ω	900/100E
C9	0,15	μF	906/150K	R29	10	Ω	900/10E
C11)	50	μF	913/P50+50	R30	100	Ω	900/100E
C11a)	50	μF		R31	47000	Ω	900/47K
C12	50	μF	910/D50	R32	47000	Ω	900/47K
C13 )	50	μF	913/P50+50	R33	100	Ω	900/100E
C13a)	50	μF		R34	10	Ω	900/10E
C14	50	μF	AC 5951/50	R35	10	Ω	900/10E
C16	39	pF	905/39E	R36	100	Ω	900/100E
C17	10.000	pF	906/10K	R37	27000	Ω	900/27K
C18 )	50	μF	913/P50+50	R38	4700	Ω	900/4K7
C18a)	50	μF		R39	1	MΩ	901/1M
C19 )	50	μF	913/P50+50	R40	6800	Ω	2x900/15K *
C19a)	50	μF		R42	1000	Ω	900/1K
C20	2200		906/V2K2	R43	10	Ω	900/10E
R1	0,27	MΩ	900/270K	R44	10	Ω	900/10E
R2	1	MΩ	900/1M				
R3	1800	Ω	900/1K8	V11	2A (117-145V)		974/2000
R4	0,27	MΩ	900/270K	V11'	1A (200-245V)		974/1000
R5	0,56	MΩ	900/560K	V12	250	mA	974/250
R6	56000	Ω	901/56K	V13	250	mA	974/250
R7	0,1	MΩ	900/100K	V14	20	A	08 143 02
R8	1	MΩ	900/1M				
R9	1000	Ω	901/1K	Gr1	390	mA	3x SR300B130
R10	56000	Ω	900/56K	Gr2	390	mA	3x SR300B130
R11	0,1	MΩ	900/100K				
R12	0,18	MΩ	901/180K				
R13	0,1	MΩ	901/100K				
R14	18000	Ω	901/18K				

DJ/GH